

TSK Fuzzy System
Dynamic TSK Fuzzy System

Closed-Loop Dynamics with Fuzzy Controller

Stability Analysis

Stable Fuzzy Controllers

PID Controller Using Fuzzy Systems

Takagi-Sugeno-Kang Fuzzy System (TSK)[1]

- ▶ A TSK fuzzy system is constructed from the following rules:
IF x_1 is C_1^I and ... and x_n is C_n^I THEN $y^I = f(x_1, \dots, x_n)$
- ▶ $y^I = f(x_1, \dots, x_n)$ is a crisp function, and can be any general fcn.
- ▶ Usually two types of TSK fuzzy system is applied
 1. Zero-Order Sugeno Model
 - ▶ y^I is const.
 - ▶ It is a special case of the product inf. , singleton fuzzifier,
 2. First-Order Sugeno Model
 - ▶ y^I is a linear fcn. of x_i
- ▶ The output of the TSK fuzzy system is computed as the weighted average of the y^I 's
$$y^* = \frac{\sum_{I=1}^M y^I w^I}{\sum_{I=1}^M w^I}$$
where $w^I = \prod_{i=1}^n \mu_{C_i^I}(x_i)$

Dynamic TSK Fuzzy System

- ▶ Output of a TSK fuzzy system appears as one of its inputs:
 IF $x(k)$ is A_1^p and ... and $x(k - n + 1)$ is A_n^p and $u(k)$ is B^p THEN
 $x^p(k + 1) = a_1^p x(k) + \dots + a_n^p x(k - n + 1) + b^p u(k)$
 - ▶ A^p and B^p are fuzzy sets
 - ▶ a^p and b^p are const., $p = 1, 2, \dots, N$,
 - ▶ $u(k)$: input to the system
 - ▶ $\mathbf{x}(k) = (x(k), x(k - 1), \dots, x(k - n + 1))^T \in R^n$: the state vector of the system.

- ▶ Output of the TSK is

$$x^*(k + 1) = \frac{\sum_{p=1}^N x^p(k + 1) v^p}{\sum_{p=1}^N v^p}$$

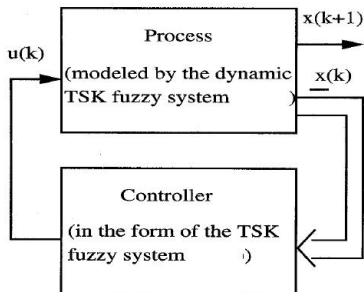
where $v^p = \prod_{i=1}^n \mu_{A_i^p}[x(k - i + 1)] \mu_{B^p}[u(k)]$

- ▶ Dynamic TSK fuzzy system can be applied to model dynamics of a plant

Closed-Loop Dynamics of Fuzzy Model with Fuzzy Controller

► Consider a feedback control system

- The process under control is modeled by the dynamic TSK fuzzy model
- The controller is the TSK fuzzy system with $c_0^l = 0$ and $x_i = x(k - i + 1)$ for $i = 1, 2, \dots, n$



- ▶ The closed-loop fuzzy control system is equivalent to the dynamic TSK fuzzy system by the following rules:
 IF $x(k)$ is $(C_1^l \text{ and } A_1^p)$ and $x(k - n + 1)$ is $(C_n^l \text{ and } A_n^p)$ THEN
 $x^{lp}(k + 1) = \sum_{i=1}^n (a_i^p + b^p c_i^l) x(k - i + 1)$
 - ▶ $u(k)$: the output of the controller,
 - ▶ $l = 1, 2, \dots, M$, $p = 1, 2, \dots, N$
 - ▶ fuzzy sets $(C_i^l \text{ and } A_i^p)$ are characterized by the mem. fcn.
 $\mu_{C_i^l}(x(k - i + l)), \mu_{A_i^p}(x(k - i + 1))$.
- ▶ The output of this dynamic TSK fuzzy system:

$$x(k + 1) = \frac{\sum_{l=1}^M \sum_{p=1}^N x^{lp}(k+1) w^l v^p}{\sum_{l=1}^M \sum_{p=1}^N w^l v^p}$$

where

- ▶ $w^l = \prod_{i=1}^n \mu_{C_i^l}(x(k - i + 1))$
- ▶ $v^p = \prod_{i=1}^n \mu_{A_i^p}[x(k - i + 1)] \mu_{B^p}[u(k)]$

Stability Analysis of the Dynamic TSK Fuzzy System

- ▶ Consider System dynamics $x(k+1) = Ax(k)$
- ▶ Based on Lyapunov theorem, this system is globally asymptotically stable iff $\exists P > 0$ s.t. $A^T P A - P < 0$

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- ▶ Now for the TSK dynamical model define:
 - ▶ $x(k) = [x(k) \dots x(k - n_1)]^T$
 - ▶ $A_p = \begin{bmatrix} a_1^p & a_2^p & \dots & a_n^p \\ 1 & 0 & \dots & 0 \\ 0 & 1 & \dots & \dots \\ \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \dots & 1 \end{bmatrix}$
 - ▶ $b^p = 0$ (consider no input $u(k)$ for the system)
 - ▶ $\therefore$ output of the systems: $x(k+1) = \frac{\sum_{p=1}^N A_p x(k) v^p}{\sum_{p=1}^N v^p}$
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- ▶ The TSK system modeled above is stable if $\exists$ a common matrix $P > 0$ s.t. $A_p^T P A_p - P < 0$ for all $p = 1, 2, \dots, N$.

Design of Stable Fuzzy Controllers for the Fuzzy Model

1. Use The following closed-loop fuzzy control system as a dynamic TSK fuzzy system.

IF $x(k)$ is $(C_1^l \text{ and } A_1^p)$ and $x(k - n + 1)$ is $(C_n^l \text{ and } A_n^p)$ THEN
 $x^{lp}(k + 1) = \sum_{i=1}^n (a_i^p + b^p c_i^l) x(k - i + 1)$

- The output:

$$x(k + 1) = \frac{\sum_{l=1}^M \sum_{p=1}^N x^{lp}(k + 1) w^l v^p}{\sum_{l=1}^M \sum_{p=1}^N w^l v^p}$$

where $w^l = \prod_{i=1}^n \mu_{C_i^l}(x(k - i + 1))$

$v^p = \prod_{i=1}^n \mu_{A_i^p}[x(k - i + 1)] \mu_{B^p}[u(k)]$

- a_i^p and b^p , and $\mu_{A_i^p}$ are known
- the controller parameters $c_i^p, \mu_{C_i^l}$ should be designed

2. Choose c_i^l and $A_{lp} =$
$$\begin{bmatrix} a_1^p + b^p c_1^l & a_2^p + b^p c_2^l & \dots & a_n^p + b^p c_n^l \\ 1 & 0 & \dots & 0 \\ 0 & 1 & \dots & \dots \\ \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \dots & 1 \end{bmatrix}$$
 where
 $l = 1, \dots, M; p = 1, \dots, N$

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 where $l = 1, \dots, M; p = 1, \dots, N$

3. Find a common matrix $P > 0$ s.t. $A_{lp}^T P A_{lp} - P < 0$ for all $p = 1, 2, \dots, N; l = 1, \dots, M$. If you could not find such P , go back to step 2 and redefine a new set of c_i^l 's

Example

► Consider a TSK dynamical system:

- IF $x(k)$ is G_1 THEN

$$x^1(k+1) = 2.18x(k) - 0.59x(k-1) - 0.603u(k)$$

- IF $x(k)$ is G_2 THEN

$$x^2(k+1) = 2.26x(k) - 0.36x(k-1) - 1.120u(k)$$

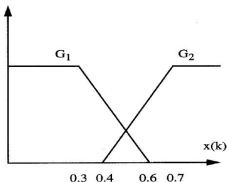
► and TSK controller:

- IF $x(k)$ is G_1 THEN

$$u^1(k) = c_1^1x(k) + c_2^1x(k-1)$$

- IF $x(k)$ is G_2 THEN

$$u^2(k) = c_1^2x(k) + c_2^2x(k-1)$$



Example Cont'd

► Step 1: design a closed loop TSK system

- IF $x(k)$ is (G_1, G_1) THEN

$$x^{11}(k+1) = (2.18 - 0.603c_1^1)x(k) + (-0.59 - 0.603c_2^1)x(k-1)$$

- IF $x(k)$ is (G_1, G_2) THEN

$$x^{12}(k+1) = (2.18 - 0.603c_1^2)x(k) + (-0.59 - 0.603c_2^2)x(k-1)$$

- IF $x(k)$ is (G_2, G_1) THEN

$$x^{21}(k+1) = (2.26 - 1.120c_1^1)x(k) + (-0.63 - 1.120c_2^1)x(k-1)$$

- IF $x(k)$ is (G_2, G_2) THEN

$$x^{22}(k+1) = (2.26 - 1.120c_1^2)x(k) + (-0.63 - 1.120c_2^2)x(k-1)$$

- Step 2: define matrices A_{lp} :

$$A_{11} = \begin{bmatrix} 2.18 - 0.603c_1^1 & -0.59 - 0.603c_2^1 \\ 1 & 0 \end{bmatrix}; A_{12} = \begin{bmatrix} 2.18 - 0.603c_1^2 & -0.59 - 0.603c_2^2 \\ 1 & 0 \end{bmatrix}; A_{21} = \begin{bmatrix} 2.26 - 1.120c_1^1 & -0.63 - 1.120c_2^1 \\ 1 & 0 \end{bmatrix}; A_{22} = \begin{bmatrix} 2.26 - 1.120c_1^2 & -0.63 - 1.120c_2^2 \\ 1 & 0 \end{bmatrix}$$

- Step 3: by trail and error proper c_i^j are:

$$c_1^1 = 1.564; c_2^1 = 0.223; c_1^2 = 0.912; c_2^2 = 0.079$$

PID Controller

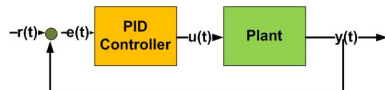
- ▶ The transfer function of a PID controller:

$$G(s) = K_p + K_i/s + K_d s$$

- ▶ $\therefore u(t) = K_p[e(t) + \frac{1}{T_i} \int_0^t e(r)dr + T_d \dot{e}(t)]$

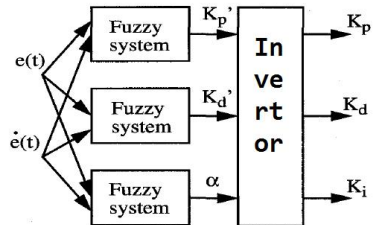
- ▶ $T_i = K_p/K_i, T_d = K_d/K_p$

- ▶ The PID gains are usually turned by experienced human experts based on some "rule of thumb."
- ▶ By using fuzzy systems, we are trying to adjust the PID gains online



Fuzzy System for PID [2]

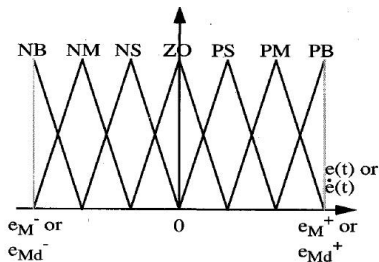
- Assume the range of PID gains:
 - $K_p \in [K_{pmax}, K_{pmin}] \subset R$
 - $K_d \in [K_{dmax}, K_{dmin}] \subset R$
- Normalize K_p and K_d to the range of $[0, 1]$
 - $K_{p'} = \frac{K_p - K_{pmin}}{K_{pmax} - K_{pmin}}$
 - $K_{d'} = \frac{K_d - K_{dmin}}{K_{dmax} - K_{dmin}}$
- Assume $T_i = \alpha T_d \rightsquigarrow K_i = \frac{K_p}{\alpha T_d} = \frac{K_p^2}{\alpha K_d}$
- Inputs of fuzzy system: $e(t), \dot{e}(t)$
- Output of fuzzy system: $K_{p'}, K_{d'}, \alpha$



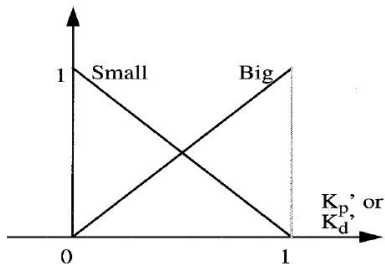
► Domain of interest:

- $e(t) \in [e_M^-, e_M^+]$
- $\dot{e}(t) \in [e_{Md}^-, e_{Md}^+]$

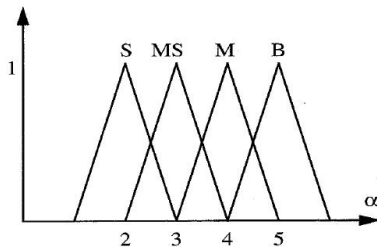
► Mem. fcn for $e(t)$ and $\dot{e}(t)$



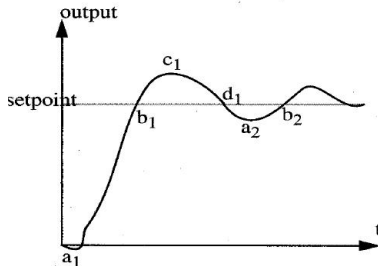
► Mem. fcn for $K_{p'}$ and $K_{d'}$



► Mem. fcn for α



- Derive the rules experimentally based on the typical step response



- For e.g., Around b_1 , a small control signal requires to avoid a large overshoot $\therefore$:
 - Small K_p
 - Large K_d
 - Small α
- IF $e(t)$ is *ZO* and $\dot{e}(t)$ is *NB*, THEN K_p is Small, K_d is Big, α is *B*
- Find the rules for other points.

Rules for K_p'

		$\dot{e}(t)$							
		NB	NM	NS	ZO	PS	PM	PB	
$e(t)$	NB	B	B	B	B	B	B	B	
	NM	S	B	B	B	B	B	S	
	NS	S	S	B	B	B	S	S	
	ZO	S	S	S	B	S	S	S	
	PS	S	S	B	B	B	S	S	
	PM	S	B	B	B	B	B	S	
	PB	B	B	B	B	B	B	B	

Rules for K_d'

		$\dot{e}(t)$							
		NB	NM	NS	ZO	PS	PM	PB	
$e(t)$	NB	S	S	S	S	S	S	S	
	NM	B	B	S	S	S	B	B	
	NS	B	B	B	S	B	B	B	
	ZO	B	B	B	B	B	B	B	
	PS	B	B	B	S	B	B	B	
	PM	B	B	S	S	S	B	B	
	PB	S	S	S	S	S	S	S	

► Rules for α

		$\dot{e}(t)$							
		NB	NM	NS	ZO	PS	PM	PB	
$e(t)$	NB	2	2	2	2	2	2	2	
	NM	3	3	2	2	2	3	3	
	NS	4	3	3	2	3	3	4	
	ZO	5	4	3	3	3	4	5	
	PS	4	3	3	2	3	3	4	
	PM	3	3	2	2	2	3	3	
	PB	2	2	2	2	2	2	2	

- There are 49 rules for each output
- Consider a fuzzy system with product inference engine, singleton fuzzifier, and center average defuzzifier

$$K_{p'} = \frac{\sum_{l=1}^{49} \bar{y}_p^l \mu_{A^l}(e(t)) \mu_{B^l}(\dot{e}(t))}{\sum_{l=1}^{49} \mu_{A^l}(e(t)) \mu_{B^l}(\dot{e}(t))} \quad K_{d'} = \frac{\sum_{l=1}^{49} \bar{y}_d^l \mu_{A^l}(e(t)) \mu_{B^l}(\dot{e}(t))}{\sum_{l=1}^{49} \mu_{A^l}(e(t)) \mu_{B^l}(\dot{e}(t))}$$

$$\alpha(t) = \frac{\sum_{l=1}^{49} \bar{y}_\alpha^l \mu_{A^l}(e(t)) \mu_{B^l}(\dot{e}(t))}{\sum_{l=1}^{49} \mu_{A^l}(e(t)) \mu_{B^l}(\dot{e}(t))}$$



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