

Nonlinear Control Lecture 7: Stability Analysis IV

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Input-to-State Stability Stability of Cascade System

Input-Output Stability

 ${\mathcal L}$ Stability of State Models

Absolute Stability Positive Real Systems Circle Criterion Popov Criterion





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Input-to-State Stability

• Consider the system $\dot{x} = f(t, x, u)$

where $f:[0,\infty) \times R^m \longrightarrow R^n$ is piecewise continuous, bounded for $\forall t \geq 0$

- Suppose the Equ. pt. of the unforced system below is **g.u.a.s**. $\dot{x} = f(t, x, 0)$
- What can be said about the behavior of the forced system in the presence of a bounded input u(t).
- For an LTI system: $\dot{x} = Ax + Bu$

where A is Hurwitz, the solution satisfies: $\|x(t)\| \leq ke^{-\lambda(t-t_0)} \|x(t_0)\| + \frac{k\|B\|}{\lambda} \sup_{t_0 \leq \tau \leq t} \|u(\tau)\|$

- Zero-input response decays to zero

Input-to-State Stability

- ► Can this conclusion be extended to nonlinear system (1)?
 - The answer in general is no, for instance:

$$\dot{x} = -3x + (1+2x^2)u$$

when u = 0, the origin is **g.e.s**

- However, with x(0) = 2 and u(t) ≡ 1, x(t) = (3 e^t)/(3 2e^t) is unbounded and have a finite scape time.
- ▶ View the system $\dot{x} = f(t, x, u)$ as a perturbation of the unforced system $\dot{x} = f(t, x, 0)$.
- Suppose there exists a Lyap. fcn for the unforced system and calculate V in the presence of u
- Since u is bounded, it may be possible to show that V is n.d. outside of a ball with radius µ where µ depends on sup ||u||.
- This is possible, for instance if the function f(t, x, u) is Lip. in u, i.e.

$$\|f(t,x,u) - f(t,x,0)\| \leq L \|u\|_{C^{\infty}}$$



- Having shown V is negative outside of a ball, ultimate boundedness theorem can be used, i.e.
 ||x(t)|| is bounded by a class KL fcn β(||x(t₀)||, t − t₀) over [t₀, t₀ + T]
 - ▶ ||x(t)|| is bounded by a class \mathcal{KL} fcn $\beta(||x(t_0)||, t t_0)$ over $[t_0, t_0 + T]$ and by a class \mathcal{K} fcn $\alpha^{-1}(\alpha_2(\mu))$ for $t \ge t_0 + T$
- ► Hence, $\|x(t)\| \leq \beta(\|x(t_0)\|, t t_0) + \alpha^{-1}(\alpha_2(\mu)) \quad \forall t \geq t_0$
- ▶ **Definition:** The system (1) is said to be **input-to-state stable** if there exist a class \mathcal{KL} fcn β and a class \mathcal{K} fcn γ s.t. for any initial state $x(t_0)$ and any bounded input u(t), the solution x(t) exists for all $t \ge t_0$ and satisfies:

$$\|x(t)\| \leq eta(\|x(t_0)\|, t-t_0) + \gamma\left(\sup_{t_0 \leq \tau \leq t} \|u(\tau)\|
ight)$$

- If u(t) converges to zero as $t \rightarrow \infty$, so does x(t).
- with $u(t) \equiv 0$, the above equation reduces to:

$$||x(t)|| \leq \beta(||x(t_0)||, t-t_0)$$

implying the origin of unforced system is g.u.a.s. $\langle B \rangle \langle \Xi \rangle \langle \Xi \rangle \langle \Xi \rangle \langle \Xi \rangle$

Sufficient condition for input-to-state stability:

Input-to-State Stability Input-Output Stability Absolute Stability

▶ Theorem: Let $V : [0, \infty) \times R^n \longrightarrow R$ be a cont. diff. fcn. s.t.

$$\begin{array}{rcl} \alpha_1(\|x\|) &\leq & V(t,x) \leq & \alpha_2(\|x\|) \\ \frac{\partial V}{\partial t} + \frac{\partial V}{\partial x} f(t,x,u) &\leq & -W_3(x), & \forall & \|x\| \geq & \rho(\|u\|) &> 0 \end{array}$$

 $\forall (t, x, u) \in [0, \infty) \times \mathbb{R}^n \times \mathbb{R}^m$ where α_1 and α_2 are class \mathcal{K}_{∞} fcns and $W_3(x)$ is a cont. p.d. fcn. on \mathbb{R}^n . Then, the system (1) is input-to-state stable with

$$\gamma = \alpha_1^{-1} \circ \alpha_2 \circ \rho$$

Input-to-State Stability Input-Output Stability Absolute Stability

- ► Lemma: Suppose f(t, x, u) is cont. diff. and globally Lip. in (x, u), uniformly in t. If the unforced system has a globally exponentially stable Equ. pt. at the origin, then the system (1) is input-to-state stable (ISS).
- Proof:
 - View the forced system as a perturbation to unforced system
 - ► The converse theorem implies that the unforced system has a Lyap. fcn satisfying the **g.e.s** conditions.
 - The perturbation terms satisfies the Lip. cond. $\forall t \ge 0$ and $\forall (x, u)$.
 - Hence, V along the trajectories of forced system (1):

$$\begin{split} \dot{V} &= \frac{\partial V}{\partial t} + \frac{\partial V}{\partial x} f(t, x, 0) + \frac{\partial V}{\partial x} [f(t, x, u) - f(t, x, 0)] \\ &\leq -c_3 \|x\|^2 + c_4 \|x\| L \|u\| = -c_3 (1 - \theta) \|x\|^2 - c_3 \theta \|x\|^2 \\ &+ c_4 \|x\| L \|u\|, \quad 0 < \theta < 1 \end{split}$$

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$$\therefore \dot{V} \leq -c_3(1-\theta) \|x\|^2 \forall \|x\| \geq \frac{c_4 L \|u\|}{c_3 \theta}, \quad \forall \ (t,x,u)$$

► The conditions of previous theorem are satisfied with:

$$\alpha_1(r) = c_1 r^2, \ \alpha_2(r) = c_2 r^2, \ \rho(r) = (c_4 L/c_3 \theta) r$$

• ... The system is input-to-state stable with $\gamma(r) = \sqrt{c_2/c_1}(c_4L/c_3\theta)r$

The previous lemma relies on globally Lip. fcn f and global exponential stability of the origin of the unforced system for ISS.



• $\dot{x} = -3x + (1 + x^2)u$ does not satisfy the global Lip. cond.

• Example 7.1: $\dot{x} = -\frac{x}{1+x^2} + u = f(x, u)$

has a globally Lip. f since $\frac{\partial f}{\partial x} = -\frac{1-x^2}{(1+x^2)^2}$ and $\frac{\partial f}{\partial u} = 1$ and are globally bounded.

- ► The origin of unforced system $\dot{x} = -\frac{x}{1+x^2}$ is $g.a.s(V = x^2/2 \implies \dot{v} = -\frac{x^2}{1+x^2}$ n.d. for all x)
- The system is locally **e.s** because of the linearized system $\dot{x} = -x$
- ► However, the system is not **g.e.s**

$$u \equiv 1, f(x, u) \geq 1/2 \implies x(t) \geq x(t_0) + t/2 \quad \forall t \geq t_0$$

If g.e.s. and globally Lip. conds. are not satisfied, then we can use previous theorem to show ISS (i.e. find a region ||x|| ≥ μ in which V ≤ 0)



Example 7.2:
$$\dot{x} = -x^3 + u$$

has a **g.a.s.** Equ. pt. at the origin when u = 0. • Let $v = x^2/2$, then \dot{V} can be written as: $\dot{V} = -x^4 + ux = -(1-\theta)x^4 - \theta x^4 + xu$ $\leq -(1-\theta)x^4, \ \forall |x| \geq \left(\frac{|u|}{\theta}\right)^{1/3} \quad 0 < \theta < 1.$

• The system is ISS with $\gamma(r) = (r/\theta)^{1/3}$

• Example 7.3: $\dot{x} = f(x, u) = -x - 2x^3 + (1 + x^2)u^2$

has a **g.e.s.** Equ. pt. at the origin when u = 0.

► However, f is not globally Lip. Let $v = x^2/2$, then: $\dot{V} = -x^2 - 2x^4 + x(1+x^2)u^2 \leq -x^4$, $\forall |x| \geq u^2$

• The system is ISS with $\gamma(r) = r^2$

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Stability of Cascade System

• Consider the cascade system

Input-to-State Stability

$$\dot{x}_1 = f_1(t, x_1, x_2), f_1 : [0, \infty) \times R^{n_1} \times R^{n_2} \longrightarrow R^{n_1}$$
 (3)

$$\dot{x}_2 = f_2(t, x_2), f_2: [0, \infty) \times \mathbb{R}^{n_2} \longrightarrow \mathbb{R}^{n_2}$$

$$\tag{4}$$

where f_1 and f_2 are p.c. in t and locally Lip. in $x = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$.

- Suppose $\dot{x}_1 = f_1(t, x_1, 0)$ and (4) both have **g.u.a.s.** Equ. pt. at $x_1 = 0$ and $x_2 = 0$.
- Under what condition the origin of the cascade system is also g.u.a.s.?
- The condition is that (3) should be **ISS** with x_2 viewed as input.
- Lemma: Under the assumption given above, if the system (3) with x₂ as input, is ISS and the origin of (4) is g.u.a.s., then the origin of the cascade system (3) and (4) is g.u.a.s.



- The foundation of input-output (I/O) approaches to nonlinear systems can be found in 1960's by Sandberg and Zames
- An input-output model relates output to input with no knowledge of the internal structure (state equation).

$$y = Hu, \quad u: [0,\infty) \to \mathcal{R}^m$$

• The norm function ||u|| should satisfy the three properties

- 1. ||u|| = 0 iff u = 0 and it is strictly positive otherwise
- 2. scaling property $\forall a > 0, u \Rightarrow ||au|| = a||u||$
- 3. triangular inequality: $\forall u_1, u_2, \|u_1 + u_2\| \le \|u_1\| + \|u_2\|$

► Example: $||u||_{\mathcal{L}_{\infty}^{m}} = \sup_{t \ge 0} ||u|| < \infty$

$$\begin{aligned} \|u\|_{\mathcal{L}_{2}^{m}} &= \sqrt{\int_{0}^{\infty} u^{T}(t)u(t)dt} < \infty \\ \|u\|_{\mathcal{L}_{p}^{m}} &= \int_{0}^{\infty} \|u\|^{p}dt)^{1/p} < \infty, \quad 1 \leq p < \infty \end{aligned}$$

- Stable system: any "well-behaved" input generate a "well-behaved" output
- Extended space: $\mathcal{L}_e^m = \{ u | u_\tau \in \mathcal{L}^m, \forall \tau \in [0, \infty) \}$

Input-Output Stability

- where u_{τ} is a truncation of u: $u_{\tau}(t) = \begin{cases} u(t) & 0 \leq t \leq \tau \\ 0 & t > \tau \end{cases}$
- It allows us to deal with unbounded "ever-growing" signals
- ▶ Example: $u(t) = t \notin \mathcal{L}_{\infty}$ but $u_{\tau}(t) \in \mathcal{L}_{\infty e}$
- Casuality: mapping H : L_e^m → L_e^q is causal if the output (Hu)(t) at any time t depends only on the value of the input up to time t

$$(Hu)_{\tau} = (Hu_{\tau})_{\tau}$$

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▶ Definition: A mapping $H : \mathcal{L}_e^m \to \mathcal{L}_e^q$ is \mathcal{L} stable if there exist a class \mathcal{K} function α , defined on $[0, \infty)$ and a nonneg const. β s.t.

$$\|(Hu)_{\tau}\|_{\mathcal{L}} \leq \alpha(\|u_{\tau}\|_{\mathcal{L}}) + \beta, \quad \forall u \in \mathcal{L}_{e}^{m}, \tau \in [0, \infty)$$
(5)

It is finite-gain \mathcal{L} stable if there exist nonneg. const. γ and β s.t.

$$\|(Hu)_{\tau}\|_{\mathcal{L}} \leq \gamma(\|u_{\tau}\|_{\mathcal{L}}) + \beta, \quad \forall u \in \mathcal{L}_{e}^{m}, \tau \in [0, \infty)$$
(6)

- ▶ β is bias term \rightsquigarrow allows Hu does not vanish at u = 0
- In finite-gain \mathcal{L} stability, the smallest possible γ is desired to satisfy (6)
- \mathcal{L}_{∞} stability is bounded-input-bounded-output stability.

Example 7.4:

$$y(t) = h(u) = a + b \tanh cu = a + b \frac{e^{cu} - e^{-cu}}{e^{cu} + e^{-cu}}$$
, for $a, b, c \ge 0$

- ▶ using the fact: $\dot{h}(u) = \frac{4bc}{(e^{cu} + e^{-cu})^2} \leq bc, \forall u \in R$
- $\blacktriangleright \therefore |h(u)| \le a + bc|u|, \ \forall \ u \in R$
- ► it is finite gain \mathcal{L}_{∞} stable with $\gamma = bc$, $\beta = a \rightarrow \langle B \rangle \langle B \rangle \langle B \rangle$ is $\beta = a \rightarrow \langle B \rangle \langle B \rangle$

• Example 7.5: $y(t) = h(u) = u^2$

- $\sup_{t\geq 0} |h(u(t))| \leq (\sup_{t\geq 0} |u(t)|)^2$
- \therefore it is \mathcal{L}_{∞} stable with $\beta = 0, \ \alpha(r) = r^2$
- ▶ But it is **not** finite-gain \mathcal{L}_{∞} stable since h(u) can not be bounded by a straight line of the form $|h(u)| \leq \gamma |u| + \beta$ for all $u \in R$

► **Example 7.6**: *y* = tan *u*

- ▶ y(t) is defined only for $|u(t)| < \frac{\pi}{2}, \forall t \ge 0$ → it is not \mathcal{L}_{∞} stable
- If we restrict $|u| \le r \le \frac{\pi}{2} \rightsquigarrow |y| \le (\frac{\tan r}{r})|u|$
- it is small-gain $\mathcal L$ stable
- ▶ Definition: mapping $H : \mathcal{L}_e^m \to \mathcal{L}_e^q$ is small-signal \mathcal{L} stable/small-signal finite-gain \mathcal{L} stable if there exist r s.t. inequality (5)/(6) is satisfied for all $u \in \mathcal{L}_e^m$ with $\sup_{0 \le t \le \tau} ||u|| \le r$

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${\mathcal L}$ Stability of State Models

Input-Output Stability

- What can we say about I/O stability based o the formalism of Lyapunov stability?
- Consider

$$\dot{x} = f(t, x, u)$$

$$y = h(t, x, u)$$

$$(7)$$

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- ▶ Theorem: Consider the system (7) and take $r_u, r > 0$ s.t. $\{||x|| \le r\} \subset D$ and $\{||u|| \le r_u\} \subset D_u$. Suppose that
 - ▶ x = 0 is an e.s. Equ. point of $\dot{x} = f(t, x, 0)$ and there is a Lyap. fcn V(t, x) and positive const c_i , i = 1, ..., 4 that

$$c_1 ||x||^2 \leq V(t,x) \leq c_2 ||x||^2, \frac{\partial V}{\partial t} + \frac{\partial V}{\partial x} f(t,x) \leq -c_3 ||x||^2$$
$$\frac{\partial V}{\partial x} \leq c_4 ||x|| \qquad \forall (t,x) \in [0,\infty) \times D,$$

► $\forall (t, x, u) \in [0, \infty) \times D \times D_u$ and for some nonneg. const. L, η_1 , and η_2 : $\|f(t, x, u) - f(t, x, 0)\| \le L \|u\|$, $\|h(t, x, u)\| \le \eta_1 \|x\| + \eta_2 \|u\|$

► Then for each $||x_0|| \le r\sqrt{c_1/c_2}$ the system is small-signal finite-gain \mathcal{L}_p stable for each $p \in [1, \infty]$. In particular, for each $u \in \mathcal{L}_{pe}$ with $\sup_{0 \le t \le \tau} ||u|| \le \min\{r_u, c_1c_3r/(c_2c_4L)\}$ the output satisfies: $||y_\tau||_{\mathcal{L}_q} \le \gamma ||u_\tau||_{\mathcal{L}_p} + \beta, \quad \tau \in [0, \infty)$

$$\gamma = \eta_2 + \frac{\eta_1 c_2 c_4 L}{c_1 c_3}, \ \beta = \eta_1 \| x_0 \| \sqrt{\frac{c_2}{c_1}} \rho, \ \rho = \begin{cases} 1, & p = \infty \\ \left(\frac{2c_2}{c_3 p}\right)^{1/\rho}, & p \in [1, \infty) \end{cases}$$



- $\mathcal L$ Stability of State Models
- Theorem Cont'd. If the origin is g.e.s and all assumptions hold for globally (with D = Rⁿ and D_u = R^m), then for each x₀ ∈ Rⁿ the system is finite-gain L_p stable for each p ∈ [1,∞]
- Exercise: Provide similar conditions for finite-gain L_p stability of LTI system

$$\dot{x} = Ax + Bu y = Cx + Du$$

► Example 7.7: $\dot{x} = -x - x^3 + u, \quad x(0) = x_0$ y = tanhx + u

• The origin of $\dot{x} = -x - x^3$ is g.e.s. (Use Lyap $V(x) = x^2/2$)

•
$$c_1 = c_2 = 1/2, \ c_3 = c_4 = 1$$

- $L = \eta_1 = \eta_2 = 1$
- The system is finite-gain \mathcal{L}_p stable

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\mathcal{L} Stability of State Models

► Example 7.8: $\dot{x}_1 = -x_2$ $\dot{x}_2 = -x_1 - x_2 - atanhx_1 + u$, $a \ge 0$ $y = x_1$

Input-Output Stability Absolute Stability

- For unforced system, take $V(x) = x^T P x = p_{11}x_1^2 + 2p_{12}x_1x_2 + p_{22}x_2^2$
- $V = -2p_{12}(x_1^2 + ax_1 tanhx_1) + 2(p_{11} p_{12} p_{22})x_1x_2 2ap_{22}x_2 tanhx_1 2(p_{22} p_{12})x_2^2$
- To cancel the cross product term x_1x_2 , choose $p_{11} = p_{12} + p_{22}$
- To make *P* p.d., choose $p_{22} = 2p_{12} = 1$
- Use the facts: $x_1 \tanh x_1 > 0, \forall x_1 \in R, |x_1| \ge |\tanh x_1|$, and $x_1^2 + x_2^2 \ge 2a|x_1||x_2|$
- $\rightarrow V = -x_1^2 x_2^2 ax_1 \tanh x_1 2ax_2 \tanh x_1 \le -\|x\|_2^2 + 2ax|x_1||x_2|$
- : for all a < 1, V < 0
- $c_1 = \lambda_{min}(P), \ c_2 = \lambda_{max}(P), \ c_3 = 1 a \text{ and } c_4 = 2 ||P||_2 = 2\lambda_{max}(P)$
- $L = \eta_1 = 1, \ \eta_2 = 0$
- ▶ All conditions are satisfied globally \rightsquigarrow system is finite-gain \mathcal{L}_P stable

$\mathcal L$ Stability of State Models

Input-Output Stability

- Theorem: Consider the system (7) and take r_u, r > 0 s.t. {||x|| ≤ r} ⊂ D and {||u|| ≤ r_u} ⊂ D_u. Suppose that
 - x = 0 is an a.s. Equ. point of $\dot{x} = f(t, x, 0)$ and there is a Lyap. fcn V(t, x) and class \mathcal{K} fcns α_i , i = 1, ..., 4 that

$$\begin{array}{rcl} \alpha_1(\|x\|) &\leq & V(t,x) &\leq & \alpha_2(\|x\|), \frac{\partial V}{\partial t} + \frac{\partial V}{\partial x}f(t,x) &\leq & -\alpha_3(\|x\|) \\ & & \frac{\partial V}{\partial x} &\leq & \alpha_4(\|x\|) & & \forall (t,x) \in & [0,\infty) \times & D, \end{array}$$

- ► $\forall (t, x, u) \in [0, \infty) \times D \times D_u$ and for some class $\mathcal{K} \alpha_i$, i = 5, ..., 7, nonneg conts. η : $\|f(t, x, u) - f(t, x, 0)\| \le \alpha_5(\|u\|)$, $\|h(t, x, u)\| \le \alpha_6(\|x\|) + \alpha_7(\|u\|) + \eta$
- Then for each $||x_0|| \le \alpha_2^{-1}(\alpha_1(r))$ the system is small-signal finite-gain \mathcal{L}_{∞} stable

$\mathcal L$ Stability of State Models

- Theorem: Consider the system (7) with $D = R^n$ and $D_u = R^m$. Suppose that
 - The system is ISS.
 - ► for all $(t, x, u) \in [0, \infty) \times \mathbb{R}^n \times \mathbb{R}^m$, some class \mathcal{K} fcns α_1, α_2 and a const. $\eta \ge 0$ $\|h(t, x, u)\| \le \alpha_6(\|x\|) + \alpha_7(\|u\|) + \eta$
- Then for each $x_0 \in \mathbb{R}^n$, the system is \mathcal{L}_{∞} stable.

Input-Output Stability

► Example 7.9: $\dot{x} = -x - 2x^3 + (1 + x^2)u^2$ y = x + u

•
$$V = x^2/2 \rightsquigarrow \dot{V} = -x^2 - 2x^4 + x(1+x^2)u^2 \le -x^4 \quad \forall |x| \ge u^2$$

• : the system is ISS with $\gamma = r^2$

•
$$\alpha_6 = r^2$$
, $\alpha_7 = r$ and $\eta = 0$

 \blacktriangleright Therefore the system is \mathcal{L}_∞ stable

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\mathcal{L} Stability of State Models

Example 7.10:

$$\begin{aligned} \dot{x}_1 &= -x_1^3 + g(t)x_2 \\ \dot{x}_2 &= -g(t)x_1 - x_2^3 + u \\ y &= x_1 + x_2 \end{aligned}$$

g(t) is continuous and bounded for $t \ge 0$

Input-Output Stability

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Absolute Stability

- Many nonlinear systems are composed of a nonlinear element in feedback connection with a linear system
- Examples include actuator/sensor saturation, relay, and other hard nonlinearities.
- If the transfer fcn of the linear subsystem is positive real, (PR), it may lead to generation of a Lyap. fcn for the whole system.
- After introducing positive real (PR) and strictly positive real (SPR) functions, some frequency-domain sufficient conditions will be discussed for absolute stability in the form of SPR of some transfer functions.



Feedback connection.



Outline Input-to-State Stability Input-Output Stability Absolute Stability

Positive Real Systems

• Consider the *n*th-order SISO system:

$$G(s) = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_0}, \quad n \ge m$$

• Definition: A transfer function G(s) is **positive real**, if

 $Re[G(s)] \ge 0$ for all $Re[s] \ge 0$

it is strictly positive real if $G(s - \epsilon)$ is positive for some $\epsilon > 0$

The condition means that G(s) maps every point in RHP of s-plane (including the jω axis) to RHP of the G(s) plane.

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Positive Real Systems

Example 7.11:

$$\begin{array}{lll} G(s) & = & \displaystyle \frac{1}{s+\lambda}, & \lambda > 0 \\ G(\sigma+j\omega) & = & \displaystyle \frac{1}{\sigma+j\omega+\lambda} = \displaystyle \frac{\sigma+\lambda-j\omega}{(\sigma+\lambda)^2+\omega^2} \end{array}$$

- $\therefore Re[G(s)] \ge 0$ if $\sigma \ge 0 \implies G(s)$ is positive real.
- It is also SPR, e.g. by selecting $\epsilon = \lambda/2$

▶ It is not always easy to use the definition for higher order systems.

- ▶ **Theorem:** A transfer function G(s) is strictly positive real iff
 - 1. G(s) is a strictly stable transfer function
 - 2. The real part of G(s) is strictly positive along the j ω axis, i.e $\forall \omega \ge 0$ $Re[G(j\omega)] > 0$

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Positive Real Systems

- The theorem implies some necessary conditions:
 - G(s) is a strictly stable
 - \blacktriangleright The Nyq. plot lies entirely in RHP, the phase is always less than 90°
 - G(s) has relative degree 0 or 1.
 - ► *G*(*s*) is strictly minimum phase (all zeros are strictly in LHP)

$$G_{1}(s) = \frac{s-1}{s^{2}+as+b}$$

$$G_{2}(s) = \frac{s+1}{s^{2}-s+1}$$

$$G_{3}(s) = \frac{1}{s^{2}+as+b}$$

$$G_{4}(s) = \frac{s+1}{s^{2}+s+1}$$

► G₁ is not SPR since it is nonminimum phase, G₂ is not SPR since it is unstable, G₃ is not SPR since its relative degree is 2.

Outline Input-to-State Stability Input-Output Stability Absolute Stability

Positive Real Systems

$$G_4(j\omega) = \frac{j\omega + 1}{-\omega^2 + j\omega + 1}$$
$$Re[G_4(j\omega)] = \frac{1}{(1 - \omega^2)^2 + \omega^2} > 0 \quad \forall \omega$$

which shows that G_4 is SPR since it is also strictly stable.

The basic difference between PR and SPR functions is that PR fcns can have poles on the jω axis while SPR fcns cannot. For instance

$$G(s) = \frac{1}{s} = \frac{\sigma - j\omega}{\sigma^2 + \omega^2}$$

is clearly PR but not SPR.

- ▶ **Theorem:** A transfer function G(s) is positive real iff
 - 1. G(s) is a stable transfer function

Positive Real Transfer Matrices

- ► The extension of positive real transfer functions for MIMO systems
- ► Definition: A p × p proper rational transfer function matrix G(s) is called positive real if
 - 1. Poles of all elements of G(s) are in $Re[s] \leq 0$
 - 2. $G(j\omega) + G^{T}(-j\omega)$ is positive semi-definite for any $\omega \ge 0$ s.t. $j\omega$ is not a pole of G(s).

Absolute Stability

- 3. The poles of G(s) on the $j\omega$ are simple and the associated residue matrices are positive semi-definite Hermitian.
- It is SPR if $G(s \epsilon)$ is PR for some $\epsilon > 0$

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Kalman-Yakubovich Lemma

Many variations of Kalman-Yakubovich lemma exist:

▶ Kalman-Yakubovich (Positive Real) Lemma: Let $G(s) = C(sI - A)^{-1}B + D$ be a $p \times p$ TF where (A, B) is controllable and (C, A) is observable. Then G(s) is PR if and only if there exist matrices $P = P^T > 0$, L, and W s.t. $PA + A^T P = -L^T L$ $PB = -C^T - L^T W$

Absolute Stability

$$PB = C^{T} - L^{T} V$$
$$W^{T} W = D + D^{T}$$

▶ Kalman-Yakubovich-Popov Lemma: Let $G(s) = C(sI - A)^{-1}B + D$ be a $p \times p$ TF where (A, B) is controllable and (C, A) is observable. Then G(s) is SPR if and only if there exist matrices $P = P^T > 0$, L, and W, and a positive constant ϵ s.t.

$$PA + A^{T}P = -L^{T}L - \epsilon P$$

$$PB = C^{T} - L^{T}W$$

$$W^{T}W = D + D^{T}$$

- Consider the feedback system shown in Fig. and assume r = 0.
- The unforced system is represented by

ż	=	Ax + Bu	
y	=	Cx + Du	(8)
и	=	$-\psi(t,y)$	

where

- $x \in R^n, u, y \in R^p, (A, B)$ is controllable,
- ► (A, C) is observable,
- ▶ ψ : $[0,\infty) \times R^p \longrightarrow R^p$ is a memoryless, possibly time-varying nonlinearity, p.c. in *t* and locally Lip. in *y*.
- ► Assume the feedback system u = ψ(t, Cx + Du) has a unique solution u for every (t, x) which is always the case when D = 0.
- ► Controllability and observability assumptions ensure that {*A*, *B*, *C*, *D*} is a minimal realization.



- ► The nonlinearity ψ(.,.) is required to satisfy a sector condition:
- For a SISO system, ψ : [0,∞) × R → R satisfies a sector condition if ∃ α, β, a & b (β > α, a < 0 < b) s.t.</p>

$$[\psi(t,y)-lpha y][\psi(t,y)-eta y]\leq 0, \,\,orall y \,\,\in [a,b]$$

- If it holds ∀y ∈ [-∞,∞], the sector condition holds globally.
- We also say ψ belongs to a sector [α, β], (α, β], [α, β), (α, β)

► For MIMO: $\psi(t, y) = \begin{bmatrix} \psi_1(t, y_1) \\ \vdots \\ \psi_p(t, y_p) \end{bmatrix}$ where $\psi_i(t, y)$ satisfies the above sector condition with α_i , β_i , a_i , b_i .





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Absolute Stability

► Taking
$$K_{min} = diag(\alpha_1, \ldots, \alpha_p), K_{max} = diag(\beta_1, \ldots, \beta_p)$$
, and
 $\Gamma = \{y \in R^p \mid a_i \leq y_i \leq b_i\}$, then

$$[\psi(t,y) - \mathcal{K}_{min}y]^{\mathcal{T}}[\psi(t,y) - \mathcal{K}_{max}y] \leq 0 \quad \forall t \geq 0, \ \forall \ y \in \Gamma$$

▶ Definition: A memoryless nonlinearity ψ : [0,∞) × R^p → R^P is said to satisfy the sector condition if:

$$[\psi(t,y) - \mathcal{K}_{min}y]^{\mathcal{T}}[\psi(t,y) - \mathcal{K}_{max}y] \leq 0 \quad \forall t \geq 0, \ \forall \ y \in \Gamma$$

for some real matrices K_{min} and K_{max} where $K = K_{max} - K_{min}$ is a symmetric **p.d.** matrix and the interior of Γ is connected and contains the origin. If $\Gamma = R^p$, then ψ satisfies the sector condition globally in which case ψ is said to belong to a sector $[K_{min}, K_{max}]$ or (K_{min}, K_{max}) .

► The objective is to show that x = 0 is a.s. Eq. pt. for all nonlinearities in the sector. Such system is said to be absolutely stable => <=> = → <</p>

Absolute Stability

- ▶ Definition: Consider the system (9) where ψ satisfies the sector condition. The system is absolutely stable if x = 0 is g.u.a.s for any nonlinearity in the given sector. It is absolutely stable with a finite domain if the origin is u.a.s..
- Circle Criterion:
 - Consider system (9) and let A be Hurwitz & ψ satisfies the sector condition with K_{min} = 0, i.e

$$\psi(t,y)^{\mathsf{T}}(\psi(t,y)-\mathsf{K}y) \leq 0 \ \forall t \geq 0, \ \forall y \in \ \mathsf{\Gamma} \subset \mathsf{R}^{\mathsf{p}}$$

• Let $V(x) = x^T P x$, $P = P^T > 0$ to be chosen.

$$\dot{V} = x^T (PA + A^T P) x - 2x^T PB\psi(t, y)$$

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► Since
$$-2\psi^{T}(\psi - Ky) \ge 0$$

 $\dot{V} \le x^{T}(PA + A^{T}P)x - 2x^{T}PB\psi(t, y) - 2\psi^{T}(t, y)(\psi(t, y) - Ky)$
 $= x^{T}(PA + A^{T}P)x + 2x^{T}(C^{T}K - PB)\psi - 2\psi^{T}\psi$

► Suppose, there are matrices $P = P^T > 0$ & L and a constant $\epsilon > 0$ s.t. $PA + A^T P = -L^T L - \epsilon P$, $PB = C^T K - \sqrt{2}L^T$ $\therefore \dot{V} \leq -\epsilon x^T P x - 2x^T L^T L x + 2\sqrt{2}x^T L^T \psi - 2\psi^T \psi$ $= -\epsilon x^T P x - [Lx - \sqrt{2}\psi]^T [Lx - \sqrt{2}\psi]^T$ $\leq -\epsilon x^T P x < 0$

- ▶ V is **n.d.** if we can find P, L, , and ϵ satisfying above equations
- ► This is the case iff Z(s) = I + KC(sI A)⁻¹B is SPR, according to KYP lemma. Since (A, C) is obs. ⇒ (A, KC) is obs. since K is nonsingular



- **Lemma:** Consider system (9), where A is Hurwitz, (A, B) is controllable, (A, C) is observable & ψ satisfies the sector condition globally, then the system is absolutely stable if $Z(s) = I + KC(sI - A)^{-1}B$ is SPR.
- The condition on having A Hurwitz can be removed by applying the loop transformation as shown in Fig.





► The new linear TF is:

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$$G_{T}(s) = G(s)[I + K_{min}G(s)]^{-1} \text{ or, equivalently}$$

$$\dot{x} = (A - BK_{min}C)x + Bu$$

$$y = Cx \text{ with}$$

$$v_{T}(t, y) = \psi(t, y) - K_{min}y$$

Now, if ψ satisfies the sector condition, then ψ_T satisfies the sector condition with K = K_{max} − K_{min}.

$$\begin{aligned} (\psi - K_{\min}y)^T (\psi - K_{\max}y) &\leq 0, \quad \psi = \psi_T + K_{\min}y \implies \\ \psi_T^T (\psi_T + (K_{\min} - K_{\max})) &= \psi_T^T (\psi_T - K) \leq 0 \end{aligned}$$

► If $(A - BK_{min}C)$ is Hurwitz \implies system is absolutely stable if $Z_T(s) = I + KG_T(s)$ is SPR.



Note that

$$Z_{T}(s) = I + (K_{max} - K_{min})G(s)(I + K_{min}G(s))^{-1}$$

= $(I + K_{max}G(s))(I + K_{min}G(s))^{-1}$

► Theorem: Consider the system (9) where (A, B) is cont. and (A, C) is obs. and ψ satisfies the sector condition globally. Then, the system is absolutely stable if

$$G_T(s) = G(s)[I + K_{min}G(s)]^{-1}$$

is Hurwitz and $Z_T(s) = (I + K_{max}G(s))(I + K_{min}G(s))^{-1}$ is SPR.

► The theorem is known as multivariable circle criterion.

• Consider system (9). Let G(s) is Hurwitz and

Input-to-State Stability Input-Output Stability Absolute Stability

$$\gamma_1 = \sup_{\omega \in R} \sigma_{\max}[G(j\omega)] = \sup_{\omega \in R} \|[G(j\omega)]\|$$

where σ_{max} denotes the max. singular value, γ_1 is finite since G(s) is Hurwitz.

Suppose the nonlinearity satisfies

$$\|\psi(t,y)\| \leq \gamma_2 \|y\|, \forall t \geq 0, \ \forall y \in \mathbb{R}^p.$$

then it satisfies the sector condition with $K_{min} = -\gamma_2 I$ and $K_{max} = \gamma_2 I$.

We now need to show that

$$G_{T}(s) = G(s)[I - \gamma_{2}G(s)]^{-1} \text{ is Hurwitz and}$$

$$Z_{T}(s) = (I + \gamma_{2}G(s))(I - \gamma_{2}G(s))^{-1} \text{ is } SPR_{\text{resp}}$$



Example Cont'd

► We have
$$det(G) \neq 0 \iff \sigma_{min}(G) > 0$$

 $\sigma_{max}(G^{-1}) = \frac{1}{\sigma_{min}(G)} \quad \text{if } \sigma_{min} > 0$
 $\sigma_{min}(I+G) \ge 1 - \sigma_{max}(G)$
 $\sigma_{max}(G_1G_2) \le \sigma_{max}(G_1) \sigma_{max}(G_2)$

• $\therefore \sigma_{\min}[I - \gamma_2 G(j\omega)]) \ge 1 - \gamma_1 \gamma_2$

- If γ₁γ₂ < 1 ⇒ the plot of det(I − γ₂G(jω)) does not go through encircle the origin
- ► ... By Nyq. criterion, $[I \gamma_2 G(j\omega)]^{-1}$ is Hurwitz $\implies G_T$ and Z_T are Hurwitz

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Examle Cont'd

► Now
$$Z_T(s)$$
 is SPR if
 $Z_T(j\omega) + Z_T^T(-j\omega) > 0 \quad \forall \omega \in R$
 $Z_T((j\omega) + Z_T^T(-j\omega) = [I + \gamma_2 G(j\omega)][I - \gamma_2 G(j\omega)]^{-1}$
 $+ [I - \gamma_2 G^T(-j\omega)]^{-1}[I + \gamma_2 G^T(-j\omega)]$
 $= [I - \gamma_2 G^T(-j\omega)]^{-1} (2I - 2\gamma^2 G^T(-j\omega)G(j\omega)) [I - \gamma_2 G(j\omega)]^{-1}$

► Hence,
$$Z_T(j\omega) + Z_T^T(-j\omega)$$
 is **p.d.** iff
 $\sigma_{min}[I - \gamma_2^2 G^T(-j\omega)G(j\omega)] > 0 \quad \forall \omega \in R$

$$\sigma_{\min}[I - \gamma_2^2 G^T(-j\omega)G(j\omega)] \geq 1 - \gamma_2^2 \sigma_{\max}[G^T(-j\omega)]\sigma_{\max}[G(j\omega)]$$

$$\geq 1 - \gamma_1^2 \gamma_2^2 > 0$$

► Hence, we can conclude that the system is abs. stable if $\gamma_1 \gamma_2 < 1$.



- This shows that closing the loop around a Hurwitz TF with a nonlinearity satisfying $\|\psi\| \leq \gamma_2 \|y\|$ with sufficiently small γ_2 does not destroy the stability of the system (small gain theorem)
- ▶ In scalar case (p = 1), the conditions of the theorem can be verified graphically by Nyq. plot of $G(j\omega)$.
 - The conditions of the theorem are

1.)
$$G_{T}(s) = \frac{G(s)}{1+\alpha G(s)}$$
 is Hurwitz
2.) $Z_{T}(s) = \frac{1+\beta G(s)}{1+\beta G(s)}$ is SPR

$$Z_T(s) = \frac{1+\beta G(s)}{1+\alpha G(s)}$$
 is SPR

- To verify SPR condition, $Z_T(s)$ is SPR iff $Z_T(s)$ is Hurwitz and $Re\left[\frac{1+\beta G(j\omega)}{1+\alpha G(j\omega)}\right] > 0 \quad \forall \ \omega \in R.$
- We now consider different cases:
- First case $\beta > \alpha > 0$:
 - In this case, we have:

$$Re\left[\frac{1/\beta + G(j\omega)}{1/\alpha + G(j\omega)}\right] > 0 \quad \forall \ \omega \ \in R$$



Graphical representation of the circle criterion.

- For a point q on the Nyq. plot of G(jω), 1/β + G(jω) & 1/α + G(jω) can be represented by the line connecting q to −1/β + j0 and −1/α + j0
- The real part of the ratio of two complex numbers is positive when the angle difference is less than π/2 ⇒ θ₁ − θ₂ < π/2 when q is outside D(α, β).</p>
- G_T(s) is Hurwitz, implying that Z_T(s) is Hurwitz, if from the Nyq. criterion, the Nyq. plot of G(jω) does not intersect the point −1/α + j0 and encircle it exactly m times CCW where m is the # of poles of G in open RHP
- Conditions of the theorem are satisfied if the Nyq. plot of G(jω) does not enter the disk D(α, β) and encircle it m times CCW_□, and an encircle it m times the times are satisfied if the times times the times the times times times times the times tim

- Second case: $\beta > 0$, $\alpha = 0$
 - ► The conditions of the theorem are: G(s) is Hurwitz and $Re[1 + \beta G(j\omega)] > 0, \forall \omega \in R$

$${\sf Re}[{\sf G}(j\omega)] \ > rac{-1}{eta}, \ orall \omega \ \in \ {\sf R}.$$

- ▶ ∴ Nyq. plot of G(s) lies to the right of $s = \frac{-1}{\beta}$ line
- ▶ Third case: $\alpha < 0 < \beta$
 - The conditions of the theorem are:

$$Re\left[rac{1/eta+G(j\omega)}{1/lpha+G(j\omega)}
ight] < 0 \ \forall \ \omega \ \in R$$

- ▶ ∴ Nyq. plot of G(s) must lie inside the disk $D(\alpha, \beta)$ (prove it).
 - ▶ Nyq. plot cannot encircle $\frac{-1}{\alpha} + j0 \implies G(s)$ must be Hurwitz for $G_T(s)$ to be so.



- Theorem (Circle Criterion): Consider system (9) with p = 1 and ψ satisfies the sector condition globally. The system is absolutely stable if one of the following conditions are satisfied:
 - 1. If $0 < \alpha < \beta$, the Nyq. plot of $G(j\omega)$ does not enter the disk $D(\alpha, \beta)$ and encircles it m times in CCW.
 - 2. If $0 = \alpha < \beta$, G(s) is Hurwitz and the Nyq. plot of $G(j\omega)$ lies to the right of the line $s = \frac{-1}{\beta}$.
 - 3. If $\alpha < 0 < \beta$, $\dot{G}(s)$ is Hurwitz and the Nyq. plot of $G(j\omega)$ lies inside $D(\alpha, \beta)$.
- ► Example 7.13: $G(s) = \frac{4}{(s+1)(\frac{1}{2}s+1)(\frac{1}{3}s+1)}$
 - Since G(s) is Hurwitz, we can allow α to be negative & apply the 3rd case of the circle criterion.
 - We need $D(\alpha, \beta)$ that encloses the Nyq. plot.(This is not unique).
 - Select the disk $D(-\gamma_2, \gamma_2)$ where $\gamma_2 > 0$ to be chosen
 - ► If we set $\gamma_1 = \sup |G(j\omega)|$ and then use $\gamma_2 \gamma_1 \iff 1 @ > 4 @ = 4 @ > 4 @ > 4 @ > 4 @ > 4 @ > 4 @ > 4 @ > 4 @ > 4 @ @ > 4 @ > 4$

Example 7.13 Cont'd

- Maximum occurs at $\omega = 0$ and
 - $\gamma_1 = 4 \implies \gamma_2 < 0.25$
- ► ∴ System is absolutely stable \forall nonlinearities in the sector $[-0.25 + \epsilon, 0.25 - \epsilon]$, where $\epsilon > 0$.
- By locating the center at another point, we may be able to obtain a disk that encloses the Nyq. plot more tightly.
- Let the center be at pt. 1.5 + j0. The max distance to Nyq. plot is 2.834 ↔ radius could be 2.9 to ensure D(-1/4.4, 1/1.4) encloses the Nyq. plot.
- System is absolutely stable ∀ nonlinearity in [-.227, .714].



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- Alternatively, if we restrict α = 0 and apply the 2nd criterion, the Nyq. plot lies to the right of s = −.857 ⇒ system is absolutely stable ∀ nonlinearity in [0, 1.166].
- ► For example, using a nonlinearity like a saturation, we conclude that since \$\psi\$ belongs to [0, 1] sector ⇒ system is **g.a.s.** using 2nd condition.
- ► However, it fails the 3rd criterion with [-.25, .25] or [-.227, .714] sectors.



 Feedback connection with saturation nonlinearity.
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$$G(s) = \frac{4}{(s-1)(\frac{1}{2}s+1)(\frac{1}{3}s+1)}$$

- G(s) is not Hurwitz, we must restrict α to be positive & apply the 1st case
 - The Nyq. plot must encircles the disk D(α, β) once in CCW.
 - ► A disk inside the right lobe is encircled once in CW → no good.
 - A disk in the left lobe is encircled once in CCW.
 - ► Let us locate the center at -3.2 + j0 (≈ half way between the two ends of the lobe).
 - ► Min. distance is .1688 ⇒ choosing the radius .168 ⇒ system is abs. stable ∀ nonlinearity in [.2969, .3298].





$$egin{array}{rcl} G(s)&=&rac{s+2}{(s+1)(s-1)} & ext{ and} \ \psi(y)&=&sat(y) \end{array}$$

- ψ belong to the sector [0, 1].
- Since G(s) is not Hurwitz, we must apply the 1st case of the circle criterion which requires the sector cond. to hold with a positive α
- We cannot apply this theorem to G(s) and conclude abs. stability.
- The best we can do is to show the abs. stab. with a finite domain





Example 7.15 Cont'd

- Now, ψ satisfies the sector condition with α = 1/a, & β = 1. on the interval [-a, a].
- ► Nyq. plot of G(s) must encircle the disk D(α, 1) once in CCW.
- ► Condition is satisfied with α > .5359.
- ► Choosing α = .55 ⇒ sector interval is [-1.818, 1.818] and the Disk (.55, 1) is encircled once by the Nyq. plot in the CCW ⇒ System is **abs. stable** with a finite domain.



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► To estimate ROA, we use Lyapunove analysis. Let $V(x) = x^T P x$. $\dot{x}_1 = x_2$ $\dot{x}_2 = x_1 + u$, $u = -\psi(y)$

Input-to-State Stability Input-Output Stability Absolute Stability

$$y = 2x_1 + x_2$$

The loop transformation yields:

$$u = \alpha y - \psi(y) = \psi_T(y)$$

$$\therefore \quad \dot{x} = A_T x + B(-\psi_T(y))$$

$$y = C x$$

$$A_T = \begin{bmatrix} 0 & 1 \\ -.1 & -.55 \end{bmatrix}, \ B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \ C = \begin{bmatrix} 2 & 1 \end{bmatrix}.$$

▶ and ψ_T satisfies the sector condition with $K = 1 - \alpha = .45$.

► To find V, we need to find P, L, and C that satisfies KYP lemma $\therefore \epsilon = .02, P = \begin{bmatrix} .4946 & .4834 \\ .4834 & 1.0774 \end{bmatrix}, L = \begin{bmatrix} .2946 & -.4436 \end{bmatrix}.$

where ϵ is chosen s.t. $Z_T(s - .5\epsilon)$ is **PR** and $(\frac{\epsilon}{2}I + A_T)$ is Hurwitz.



Example 7.15 Cont'd

► ∴ V = x^TPx is a Lyap. fcn for the system and

$$\Omega_c = \{x \in R^2 \mid V(x) \leq c \}$$

where

- $\begin{array}{ll} c & \leq & \min_{\{|y|=1.818\}} V(x) = .3446 \\ \text{to ensure that } \Omega_c \text{ is} \\ \text{contained in the set} \\ \{|y| & \leq & 1.818\}. \end{array}$
- \therefore taking c = .34 gives the estimate of ROA.



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$$\dot{x} = Ax + Bu y = Cx$$
 (9)
$$u = -\psi(t, y)$$

where $x \in \mathbb{R}^n$, $u, y \in \mathbb{R}^p$, (A, B) is controllable, (A, C) is observable, and $\psi: \mathbb{R}^P \longrightarrow \mathbb{R}^P \longrightarrow \psi^T(\psi - K_Y) < 0$ with $K = diag(\beta_1, ..., \beta_p)$ in $\Gamma = \{ \mathbf{v} \in \mathbb{R}^p \mid \mathbf{a}_i < \mathbf{v}_i < \mathbf{b}_i \}.$ We chose a Lure-type Lyapunov function $V(x) = x^T P x + 2\eta \int_{0}^{y} \psi(\sigma) K d\sigma, P = P^T > 0, \eta \ge 0$ to be chosen $\therefore \dot{V} = x^{T} (PA + A^{T} P) x - 2x^{T} PB\psi(y) + 2\eta \psi^{T}(y) KC[Ax - B\psi(y)]$ $\dot{V} < x^T (PA + A^T P) x - 2x^T PB\psi(y) + 2\eta\psi^T(y)KC[Ax - B\psi(y)]$ $-2\psi^{T}(\mathbf{y})(\psi(\mathbf{y})-K\mathbf{y})$ $= x^{T}(PA + A^{T}P)x - 2x^{T}(PB - C^{T}K - \eta A^{T}C^{T}K)\psi$ $-\psi^{T}(\mathbf{y})(2I + \eta KCB + \eta B^{T}C^{T}K)\psi(\mathbf{y})$

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- Chose η s.t. $2I + \eta KCB + \eta B^T C^T K \ge 0$ by setting η sufficiently small.
- ▶ Now, let $2I + \eta KCB + \eta B^T C^T K = W^T W$ and suppose $\exists P = P^T > 0$ and L and $\epsilon > 0$ s.t.

$$PA + A^T P = -L^T L - \epsilon P$$

$$PB = C^T K + \eta A^T C^T K - L^T W$$

$$\dot{V} \leq -\epsilon x^T P x - [Lx - W\psi(y)]^T [Lx - W\psi(y)]^T \leq -\epsilon x^T P x < 0$$

• Existence of *L*, *P*, and ϵ can be verified from KYP lemma iff $Z(s) = I + \eta KCB + (KC + \eta KCA)(sI - A)^{-1}B$ $= I + (1 + \eta s)KG(s)$, is **SPR.** where $Z(\infty) + Z^{T}(\infty) = W^{T}W$

Suppose η is chosen s.t. $(1 + \eta \lambda_i) \neq 0$ for all eig. of A, i.e. $-1/\eta$ is not an eig. of $A \rightsquigarrow (A, KC + \eta KCA)$ is observable since (A, c) is observable. Farzaneh Abdollahi Nonlinear Control Lecture 7 53/58



▶ Multivariable Popov Criterion: Consider the system where A is Hurwitz (A, B) is controllable, (C, A) is observable and ψ is time-invariant decentralized nonlinearity satisfying the sector condition globally with a diagonal matrix K. The, the system is absolutely stable if $\exists \eta \geq 0$ with $-1/\eta$ not an eig. of A, s.t.

$$Z(s) = I + (1 + \eta s) KG(s)$$

is SPR.

► loop transformation can be applied for Popov criterion as well

▶ when
$$p = 1 \implies$$
 choose η s.t. $Z(\infty) > 0$. $Z(s)$ is SPR iff
 $Re[1 + (1 + j\eta\omega)kG(j\omega)] > 0 \quad \forall \ \omega \ \in \ R$
 $\frac{1}{k} + Re[G(j\omega)] - \eta\omega Im[G(j\omega)] > 0 \quad \forall \ \omega \ \in \ R$ (10)

that is the plot of $Re[G(j\omega)]$ vs. $\omega Im[G(j\omega)]$ lies to the right of the line crossing the point $\frac{-1}{k} + j0$ with the slope $\frac{1}{\eta}$.



- The plot is known as Popov plot.
- For η = 0 condition
 (10) reduces to circle criterion
 - Popov criterion is weaker than circle criterion



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$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = -x_2 - h(y)$$

$$y = x_1$$

$$A = \begin{bmatrix} 0 & 1 \\ 0 & -1 \end{bmatrix} \text{ is not Hurwitz} \implies \text{ use a loop transformation:}$$

$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = -x_2 - \alpha y + \alpha y - h(y)$$

$$y = x_1$$

$$Vow, \text{ take } A = \begin{bmatrix} 0 & 1 \\ -\alpha & -1 \end{bmatrix},$$

$$\psi(y) = h(y) - \alpha y, B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, C = \begin{bmatrix} 1 & 0 \end{bmatrix}.$$

$$A = \begin{bmatrix} 0 & 1 \\ -\alpha & -1 \end{bmatrix},$$

$$\psi(y) = h(y) - \alpha y, B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, C = \begin{bmatrix} 1 & 0 \end{bmatrix}.$$

 $[0, K], K = \beta - \alpha$

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Example 7.15 cont'd

- ► $Z(s) = 1 + (1 + \eta s) KG(s), \quad G(s) = C(sI A)^{-1}B, \quad \therefore \quad Z(\infty) = 1 \quad \forall \eta.$
- ► We have $\frac{1}{k} + \frac{\alpha - \omega^2 + \eta \omega^2}{(\alpha - \omega^2)^2 + \omega^2} > 0, \quad \forall \ \omega \in R$
- Since k > 0 ⇒ for η ≥ 1 the inequality is satisfied.

Let

$$\alpha = 1 \implies \frac{1}{k} + \frac{1 - \omega^2 + \eta \omega^2}{(1 - \omega^2)^2 + \omega^2} > 0$$

- ▶ Popov plot lies to the right of any line of slope ≤ 1.
- System is abs. stable ∀ h(.) in the sector [α, β] with α sufficiently small and β sufficiently large.



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Outline Input-to-State Stability Input-Output Stability Absolute Stability

Example 7.15 cont'd

- To check the advantage of Popov criterion: $\eta = 0$
- ► the system is abs. stable if Nyquist plot of G(jω) lies to right of vertical line Re[s] = -1/k
 - \therefore k cannot be arbitrarily large
 - $\blacktriangleright \quad \frac{1}{k} + \frac{\alpha \omega^2}{(\alpha \omega^2)^2 + \omega^2} > 0 \leadsto \ k < 1 + 2\sqrt{\alpha}$
- ▶ ∴ by circle criterion the system is abs stable when h(.) is in the sector $[\alpha, 1 + +2\sqrt{\alpha} \epsilon]$ where $\alpha > 0$ and $\epsilon > 0$ is arbitrarily small

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